

DATA SHEET

For a complete data sheet, please also download:

- The IC06 74HC/HCT/HCU/HCMOS Logic Family Specifications
- The IC06 74HC/HCT/HCU/HCMOS Logic Package Information
- The IC06 74HC/HCT/HCU/HCMOS Logic Package Outlines

74HC/HCT123

Dual retriggerable monostable
multivibrator with reset

Product specification
Supersedes data of September 1993
File under Integrated Circuits, IC06

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Dual retriggerable monostable multivibrator with reset

74HC/HCT123

FEATURES

- DC triggered from active HIGH or active LOW inputs
- Retriggerable for very long pulses up to 100% duty factor
- Direct reset terminates output pulse
- Schmitt-trigger action on all inputs except for the reset input
- Output capability: standard (except for nR_{EXT}/C_{EXT})
- I_{CC} category: MSI

GENERAL DESCRIPTION

The 74HC/HCT123 are high-speed Si-gate CMOS devices and are pin compatible with low power Schottky TTL (LSTTL). They are specified in compliance with JEDEC standard no. 7A.

The 74HC/HCT123 are dual retriggerable monostable multivibrators with output pulse width control by three methods. The basic pulse time is programmed by selection of an external resistor (R_{EXT}) and capacitor (C_{EXT}). The external resistor and capacitor are normally connected as shown in Fig.6.

Once triggered, the basic output pulse width may be extended by retriggering the gated active LOW-going edge input (n $\bar{A}$) or the active HIGH-going edge input (nB). By repeating this process, the output pulse period (nQ = HIGH, n $\bar{Q}$ = LOW) can be made as long as desired. Alternatively an output delay can be terminated at any time by a LOW-going edge on input n $\bar{R}_D$, which also inhibits the triggering.

An internal connection from n $\bar{R}_D$ to the input gates makes it possible to trigger the circuit by a positive-going signal at input n $\bar{R}_D$ as shown in the function table. Figures 7 and 8 illustrate pulse control by retriggering

and early reset. The basic output pulse width is essentially determined by the values of the external timing components R_{EXT} and C_{EXT}. For pulse widths, when C_{EXT} < 10 000 pF, see Fig.9.

When C_{EXT} > 10 000 pF, the typical output pulse width is defined as:

$$t_W = 0.45 \times R_{EXT} \times C_{EXT} \text{ (typ.)},$$

where:

t_W = pulse width in ns;

R_{EXT} = external resistor in k Ω ;

C_{EXT} = external capacitor in pF.

Schmitt-trigger action in the n $\bar{A}$ and nB inputs, makes the circuit highly tolerant to slower input rise and fall times.

The '123' is identical to the '423' but can be triggered via the reset input.

QUICK REFERENCE DATA

GND = 0 V; T_{amb} = 25 °C; t_r = t_f = 6 ns

SYMBOL	PARAMETER	CONDITIONS	TYPICAL		UNIT
			HC	HCT	
t _{PHL} / t _{PLH}	propagation delay n $\bar{A}$, nB to nQ, n $\bar{Q}$ n $\bar{R}_D$ to nQ, n $\bar{Q}$	C _L = 15 pF; V _{CC} = 5 V; R _{EXT} = 5 k Ω ; C _{EXT} = 0 pF	26	26	ns
			20	23	ns
C _I	input capacitance		3.5	3.5	pF
C _{PD}	power dissipation capacitance per monostable	notes 1 and 2	54	56	pF

Notes

1. C_{PD} is used to determine the dynamic power dissipation (P_D in μ W):

$$P_D = C_{PD} \times V_{CC}^2 \times f_i + \sum (C_L \times V_{CC}^2 \times f_o) + 0.75 \times C_{EXT} \times V_{CC}^2 \times f_o + D \times 16 \times V_{CC} \text{ where:}$$

f_i = input frequency in MHz

f_o = output frequency in MHz

D = duty factor in %

C_L = output load capacitance in pF

V_{CC} = supply voltage in V

C_{EXT} = timing capacitance in pF

$\sum (C_L \times V_{CC}^2 \times f_o)$ sum of outputs

2. For HC the condition is V_I = GND to V_{CC}
For HCT the condition is V_I = GND to V_{CC} – 1.5 V

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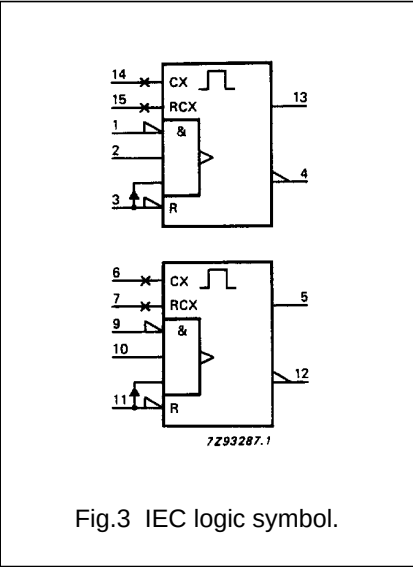
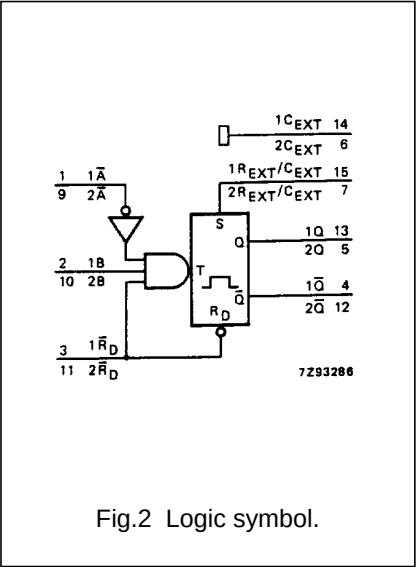
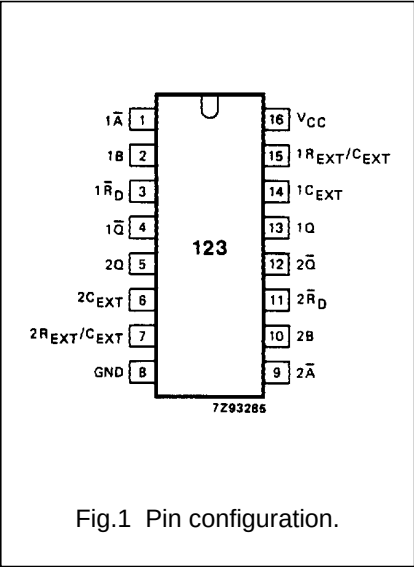
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ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
74HC123N; 74HCT123N	DIP16	plastic dual in-line package; 16 leads (300 mil); long body	SOT38-1
74HC123D; 74HCT123D	SO16	plastic small outline package; 16 leads; body width 3.9 mm	SOT109-1
74HC123DB; 74HCT123DB	SSOP16	plastic shrink small outline package; 16 leads; body width 5.3 mm	SOT338-1
74HC123PW; 74HCT123PW	TSSOP16	plastic thin shrink small outline package; 16 leads; body width 4.4 mm	SOT403-1

PIN DESCRIPTION

PIN NO.	SYMBOL	NAME AND FUNCTION
1, 9	$1\bar{A}, 2\bar{A}$	trigger inputs (negative-edge triggered)
2, 10	1B, 2B	trigger inputs (positive-edge triggered)
3, 11	$1\bar{R}_D, 2\bar{R}_D$	direct reset LOW and trigger action at positive edge
4, 12	$1\bar{Q}, 2\bar{Q}$	outputs (active LOW)
7	$2R_{EXT}/C_{EXT}$	external resistor/capacitor connection
8	GND	ground (0 V)
13, 5	1Q, 2Q	outputs (active HIGH)
14, 6	$1C_{EXT}, 2C_{EXT}$	external capacitor connection
15	$1R_{EXT}/C_{EXT}$	external resistor/capacitor connection
16	V_{CC}	positive supply voltage



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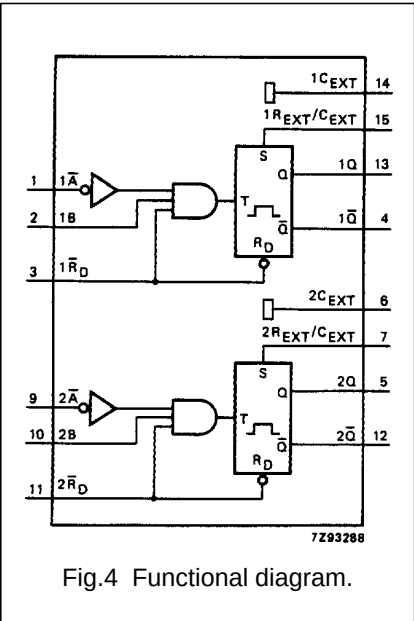


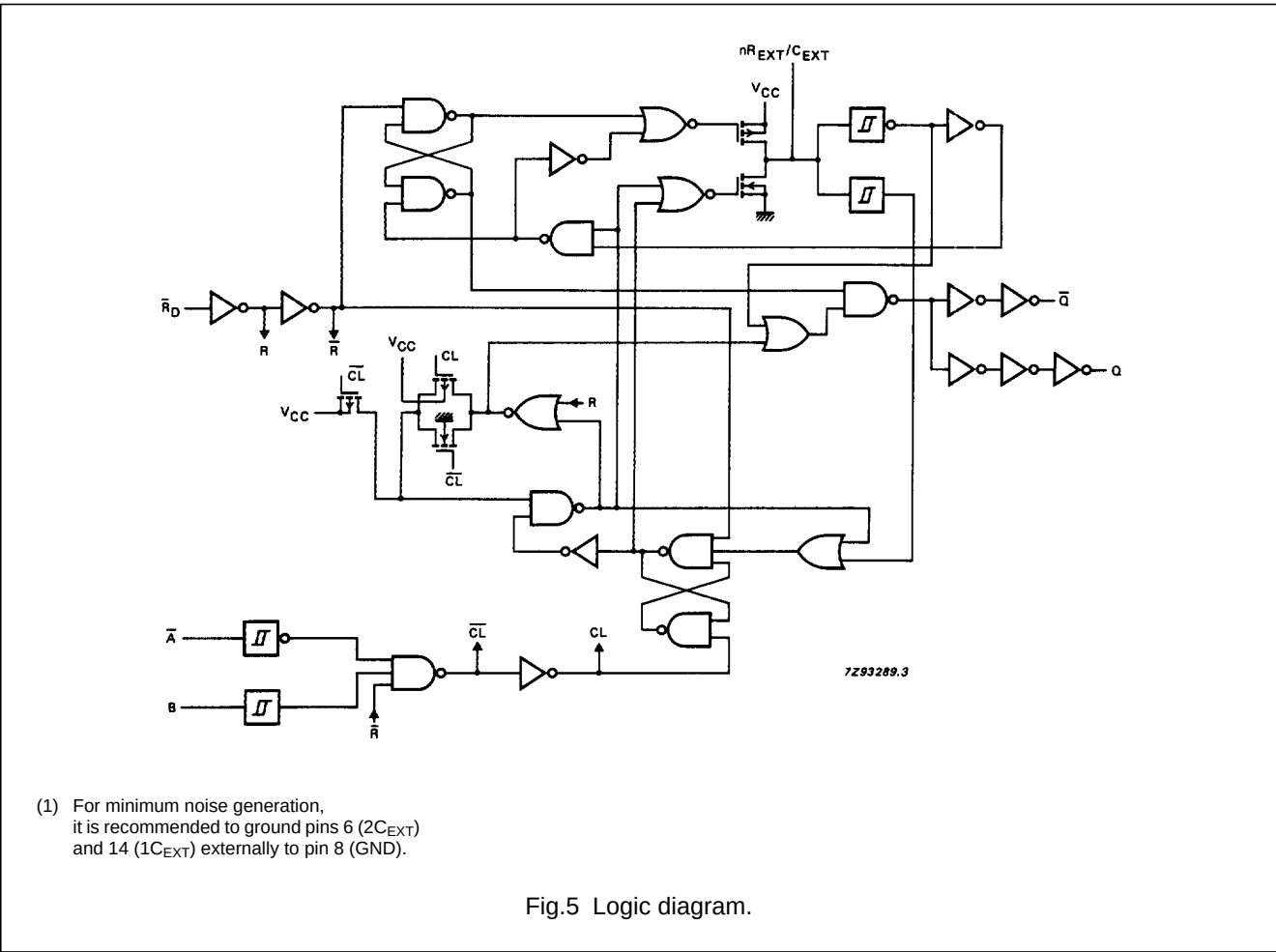
Fig.4 Functional diagram.

FUNCTION TABLE

INPUTS			OUTPUTS	
nRD	nA	nB	nQ	nQ
L	X	X	L	H
X	H	X	L ⁽¹⁾	H ⁽¹⁾
X	X	L	L ⁽¹⁾	H ⁽¹⁾
H	L	↑	[pulse]	[pulse]
H	↓	H	[pulse]	[pulse]
↑	L	H	[pulse]	[pulse]

- H = HIGH voltage level
L = LOW voltage level
X = don't care
↑ = LOW-to-HIGH transition
↓ = HIGH-to-LOW transition
[pulse] = one HIGH level output pulse
[pulse] = one LOW level output pulse

Note
1. If the monostable was triggered before this condition was established, the pulse will continue as programmed.



(1) For minimum noise generation, it is recommended to ground pins 6 (2C_EXT) and 14 (1C_EXT) externally to pin 8 (GND).

Fig.5 Logic diagram.

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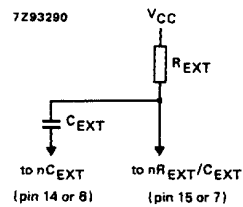


Fig.6 Timing component connections.

DC CHARACTERISTICS FOR 74HC

For the DC characteristics see *"74HC/HCT/HCU/HCMOS Logic Family Specifications"*.

Output capability: standard (except for nR_{EXT}/C_{EXT})

I_{CC} category: MSI

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AC CHARACTERISTICS FOR 74HC

GND = 0 V; $t_r = t_f = 6$ ns; $C_L = 50$ pF

SYMBOL	PARAMETER	T _{amb} (°C)							UNIT	TEST CONDITIONS		
		74HC								V _{CC} (V)	WAVEFORMS/ NOTES	
		+25			−40 to +85		−40 to +125					
		min.	typ.	max.	min.	max.	min.	max.				
t _{PLH}	propagation delay nR _D , nA, nB to nQ		83 30 24	255 51 43		320 64 54		385 77 65	ns	2.0 4.5 6.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ	
t _{PLH}	propagation delay nR _D , nA, nB to nQ		83 30 24	255 51 43		320 64 54		385 77 65	ns	2.0 4.5 6.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ	
t _{PHL}	propagation delay nR _D to nQ (reset)		66 24 19	215 43 37		270 54 46		325 65 55	ns	2.0 4.5 6.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ	
t _{PLH}	propagation delay nR _D to nQ (reset)		66 24 19	215 43 37		270 54 46		325 65 55	ns	2.0 4.5 6.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ	
t _{THL} / t _{TLH}	output transition time		19 7 6	75 15 13		95 19 16		110 22 19	ns	2.0 4.5 6.0		
t _w	trigger pulse width nA = LOW	100 20 17	8 3 2		125 25 21		150 30 26		ns	2.0 4.5 6.0	Fig.7	
t _w	trigger pulse width nB = HIGH	100 20 17	17 6 5		125 25 21		150 30 26		ns	2.0 4.5 6.0	Fig.7	
t _w	reset pulse width nR _D = LOW	100 20 17	14 5 4		125 25 21		150 30 26		ns	2.0 4.5 6.0	Fig.8	
t _w	output pulse width nQ = HIGH nQ = LOW		450		–		–		μs	5.0	C _{EXT} = 100 nF; R _{EXT} = 10 kΩ; Figs 7 and 8	
t _w	output pulse width nQ = HIGH nQ = LOW		75		–		–		ns	5.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ; note 1; Figs 7 and 8	
t _{rt}	retrigger time nA, nB		110		–		–		ns	5.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ; note 2; Fig.7	
R _{EXT}	external timing resistor	10 2		1000 1000	–		–		kΩ	2.0 5.0	Fig.9	
C _{EXT}	external timing capacitor	no limits							pF	5.0	Fig.9; note 3	

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DC CHARACTERISTICS FOR 74HCT

For the DC characteristics see *"74HC/HCT/HCU/HCMOS Logic Family Specifications"*.

Output capability: standard (except for nR_{EXT} / C_{EXT})

I_{CC} category: MSI

Note to HCT types

The value of additional quiescent supply current (ΔI_{CC}) for a unit load of 1 is given in the family specifications. To determine ΔI_{CC} per input, multiply this value by the unit load coefficient shown in the table below.

INPUT	UNIT LOAD COEFFICIENT
$\overline{nA}$, nB	0.35
$\overline{nR_D}$	0.50

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AC CHARACTERISTICS FOR 74HCTGND = 0 V; $t_r = t_f = 6$ ns; $C_L = 50$ pF

SYMBOL	PARAMETER	T _{amb} (°C)							UNIT	TEST CONDITIONS	
		74HCT								V _{CC} (V)	WAVEFORMS/ NOTES
		+25			−40 to +85		−40 to +125				
		min.	typ.	max.	min.	max.	min.	max.			
t _{PHL}	propagation delay nR _D , nA, nB to nQ̄		30	51		64		77	ns	4.5	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ
t _{PLH}	propagation delay nR _D , nĀ, nB to nQ		28	51		64		77	ns	4.5	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ
t _{PHL}	propagation delay nR _D to nQ (reset)		27	46		58		69	ns	4.5	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ
t _{PLH}	propagation delay nR _D to nQ̄ (reset)		23	46		58		69	ns	4.5	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ
t _{THL} / t _{TLH}	output transition time		7	15		19		22	ns	4.5	
t _W	trigger pulse width nA = LOW	20	3		25		30		ns	4.5	Fig.7
t _W	trigger pulse width nB = HIGH	20	5		25		30		ns	4.5	Fig.7
t _W	reset pulse width nR _D = LOW	20	7		25		30		ns	4.5	Fig.8
t _W	output pulse width nQ̄ = HIGH nQ̄ = LOW		450		–		–		μs	5.0	C _{EXT} = 100 nF; R _{EXT} = 10 kΩ; Figs 7 and 8
t _W	output pulse width nQ = HIGH nQ̄ = LOW		75		–		–		ns	5.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ; note 1; Figs 7 and 8
t _{rt}	retrigger time nĀ, nB		110		–		–		ns	5.0	C _{EXT} = 0 pF; R _{EXT} = 5 kΩ; note 2; Fig.7
R _{EXT}	external timing resistor	2		1000	–		–		kΩ	5.0	Fig.9
C _{EXT}	external timing capacitor	no limits							pF	5.0	Fig.9; note 3

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Notes to AC characteristics

1. For other R_{EXT} and C_{EXT} combinations see Fig.9.
If $C_{EXT} > 10$ nF, the next formula is valid:

$$t_W = K \times R_{EXT} \times C_{EXT} \text{ (typ.)}$$

where: t_W = output pulse width in ns;
 R_{EXT} = external resistor in k Ω ; C_{EXT} = external capacitor in pF;
 K = constant = 0.55 for $V_{CC} = 5.0$ V and 0.48 for $V_{CC} = 2.0$ V.

The inherent test jig and pin capacitance at pins 15 and 7 (nR_{EXT} / C_{EXT}) is approximately 7 pF.

2. The time to retrigger the monostable multivibrator depends on the values of R_{EXT} and C_{EXT} .
The output pulse width will only be extended when the time between the active-going edges of the trigger input pulses meets the minimum retrigger time.
If $C_{EXT} > 10$ pF, the next formula (at $V_{CC} = 5.0$ V) for the set-up time of a retrigger pulse is valid:

$$t_{rt} = 30 + 0.19 \times R_{EXT} \times C_{EXT}^{0.9} + 13 \times R_{EXT}^{1.05} \text{ (typ.)}$$

where: t_{rt} = retrigger time in ns;
 C_{EXT} = external capacitor in pF;
 R_{EXT} = external resistor in k Ω .

The inherent test jig and pin capacitance at pins 15 and 7 (nR_{EXT} / C_{EXT}) is 7 pF.

3. When the device is powered-up, initiate the device via a reset pulse, when $C_{EXT} < 50$ pF.

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AC WAVEFORMS

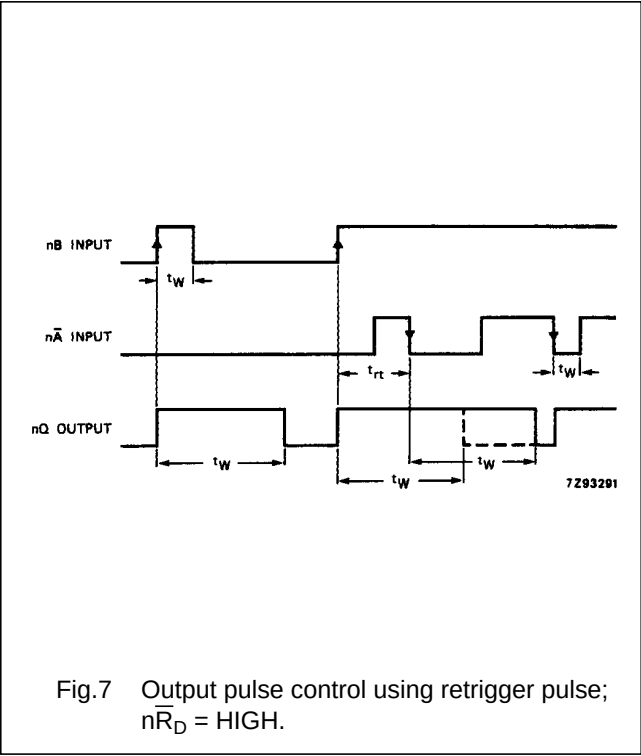


Fig.7 Output pulse control using retrigger pulse;
 $n\bar{R}_D = \text{HIGH}$.

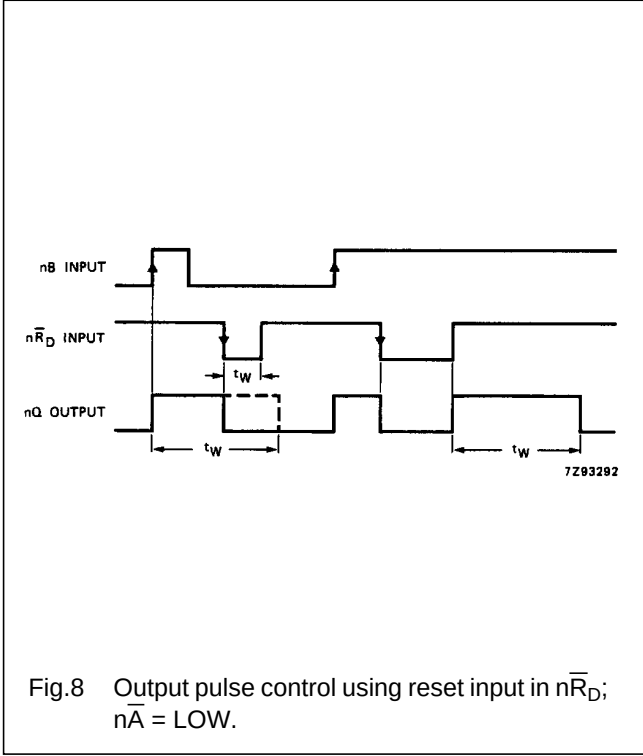


Fig.8 Output pulse control using reset input in $n\bar{R}_D$;
 $n\bar{A} = \text{LOW}$.

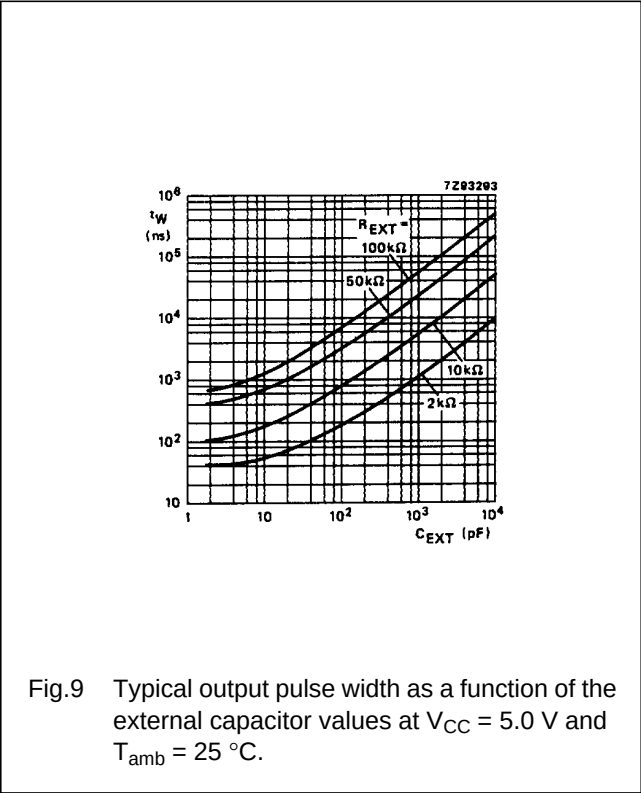


Fig.9 Typical output pulse width as a function of the
external capacitor values at $V_{CC} = 5.0\text{ V}$ and
 $T_{amb} = 25\text{ }^\circ\text{C}$.

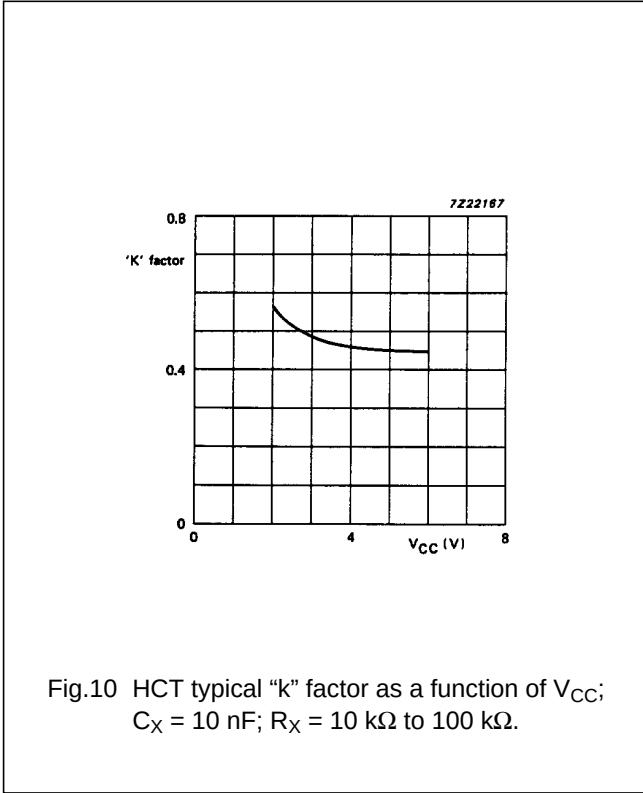


Fig.10 HCT typical "k" factor as a function of V_{CC} ;
 $C_X = 10\text{ nF}$; $R_X = 10\text{ k}\Omega$ to $100\text{ k}\Omega$.

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APPLICATION INFORMATION

Power-up considerations

When the monostable is powered-up it may produce an output pulse, with a pulse width defined by the values of R_X and C_X , this output pulse can be eliminated using the circuit shown in Fig.11.

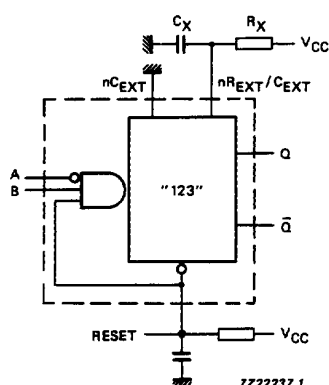


Fig.11 Power-up output pulse elimination circuit.

Power-down considerations

A large capacitor (C_X) may cause problems when powering-down the monostable due to the energy stored in this capacitor. When a system containing this device is powered-down or a rapid decrease of V_{CC} to zero occurs, the monostable may sustain damage, due to the capacitor discharging through the input protection diodes. To avoid this possibility, use a damping diode (D_X) preferably a germanium or Schottky type diode able to withstand large current surges and connect as shown in Fig.12

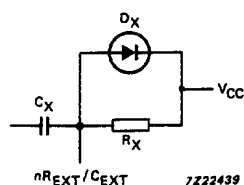


Fig.12 Power-down protection circuit.

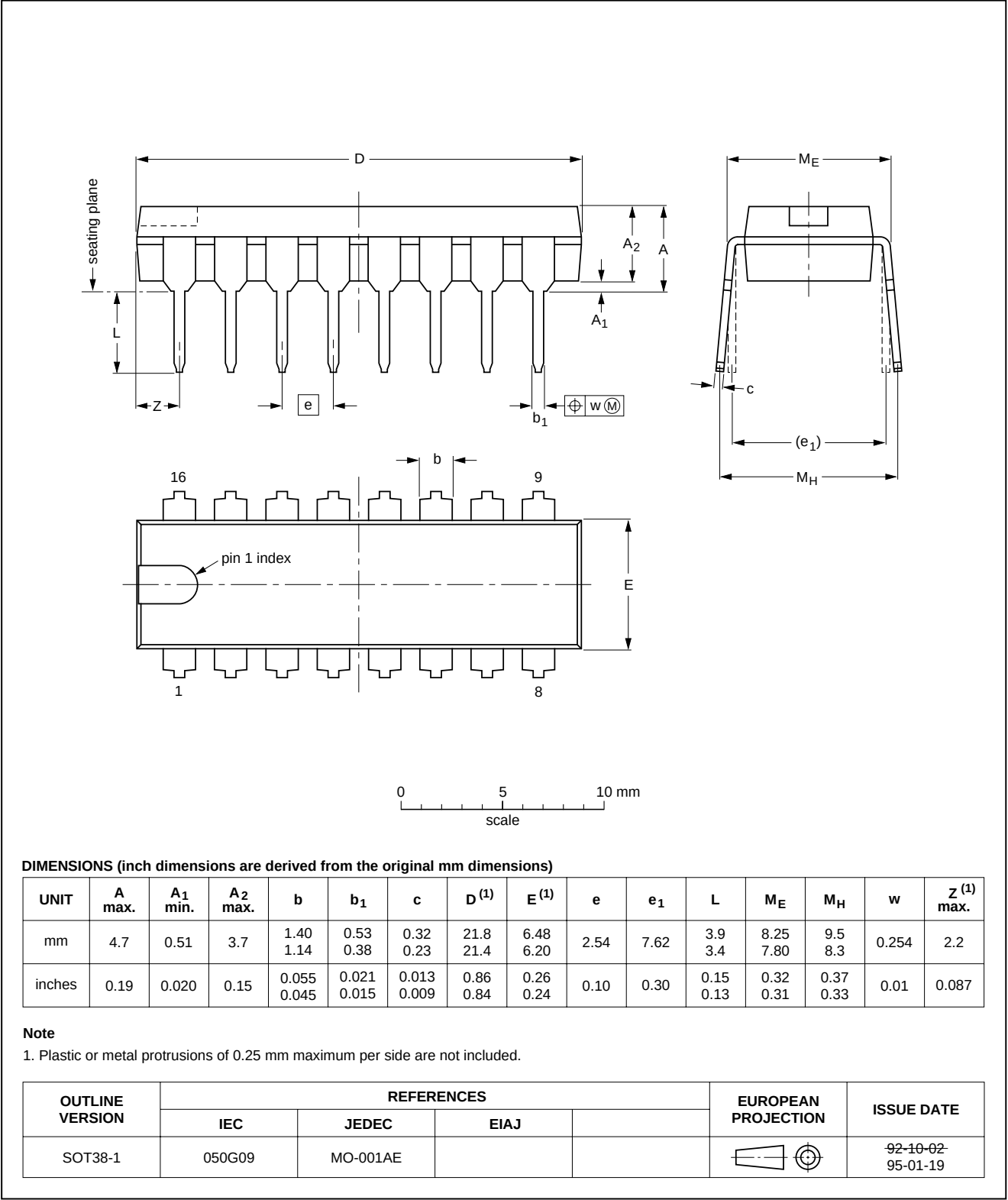
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PACKAGE OUTLINES

DIP16: plastic dual in-line package; 16 leads (300 mil); long body

SOT38-1

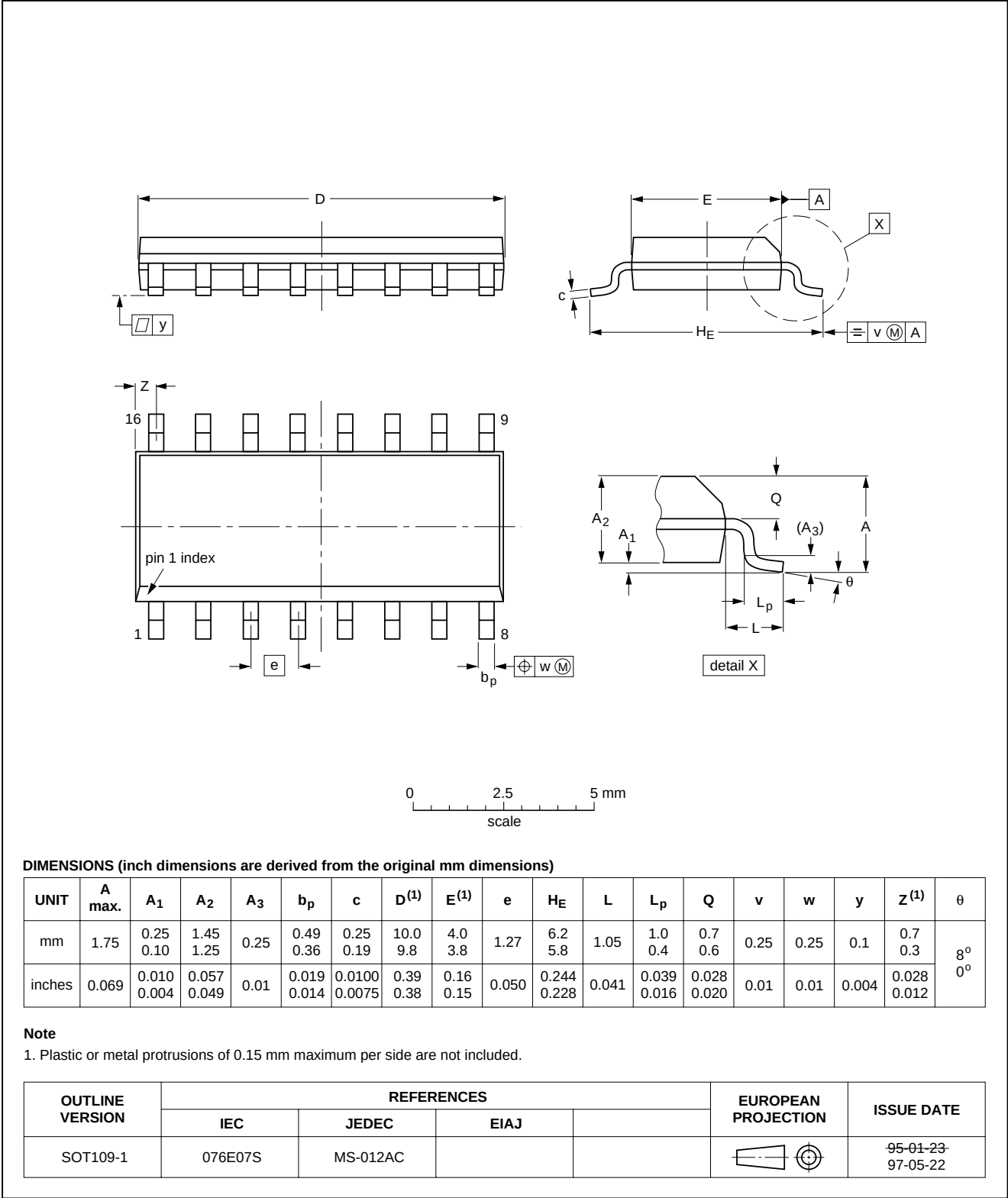


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SO16: plastic small outline package; 16 leads; body width 3.9 mm

SOT109-1

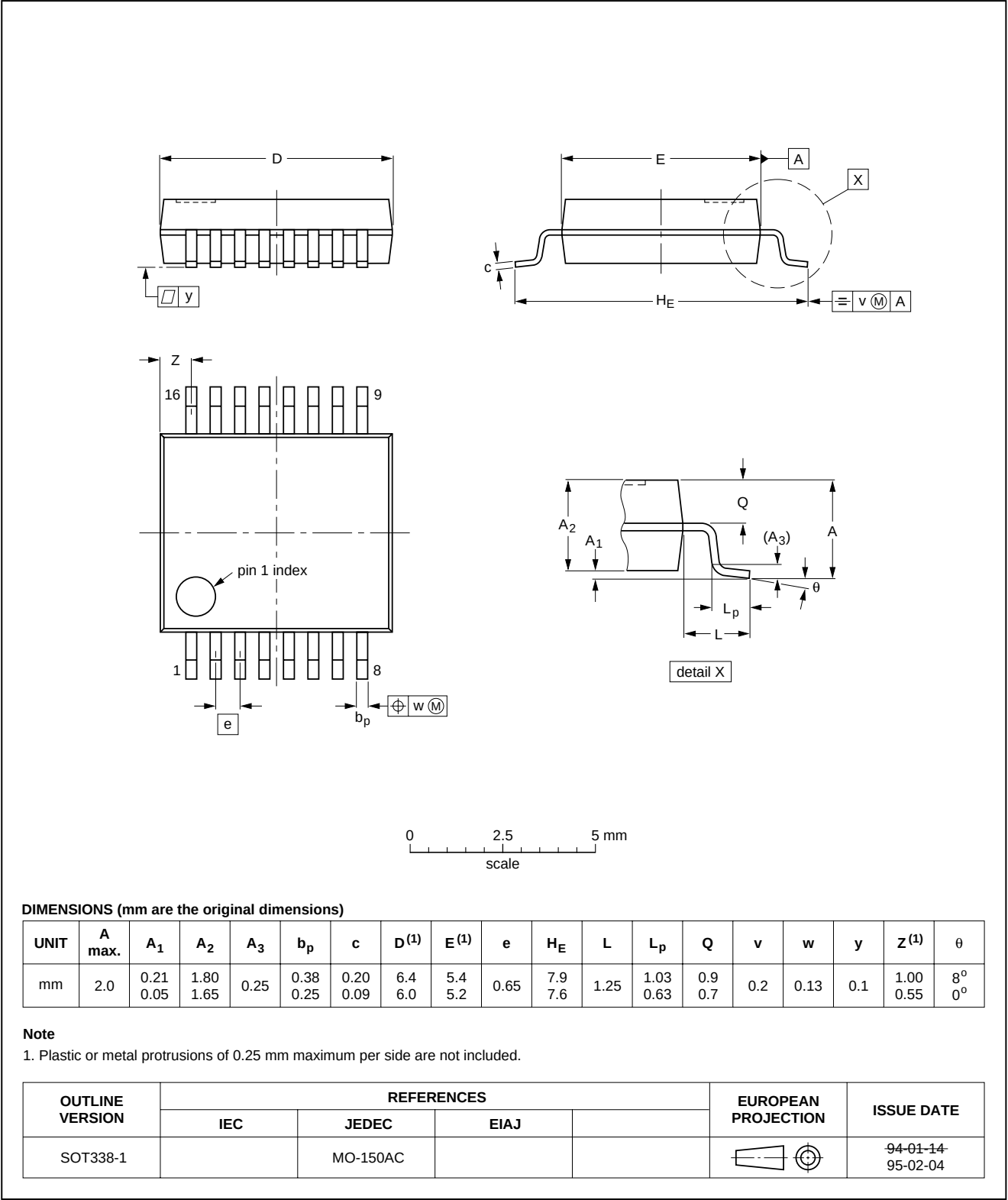


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SSOP16: plastic shrink small outline package; 16 leads; body width 5.3 mm

SOT338-1

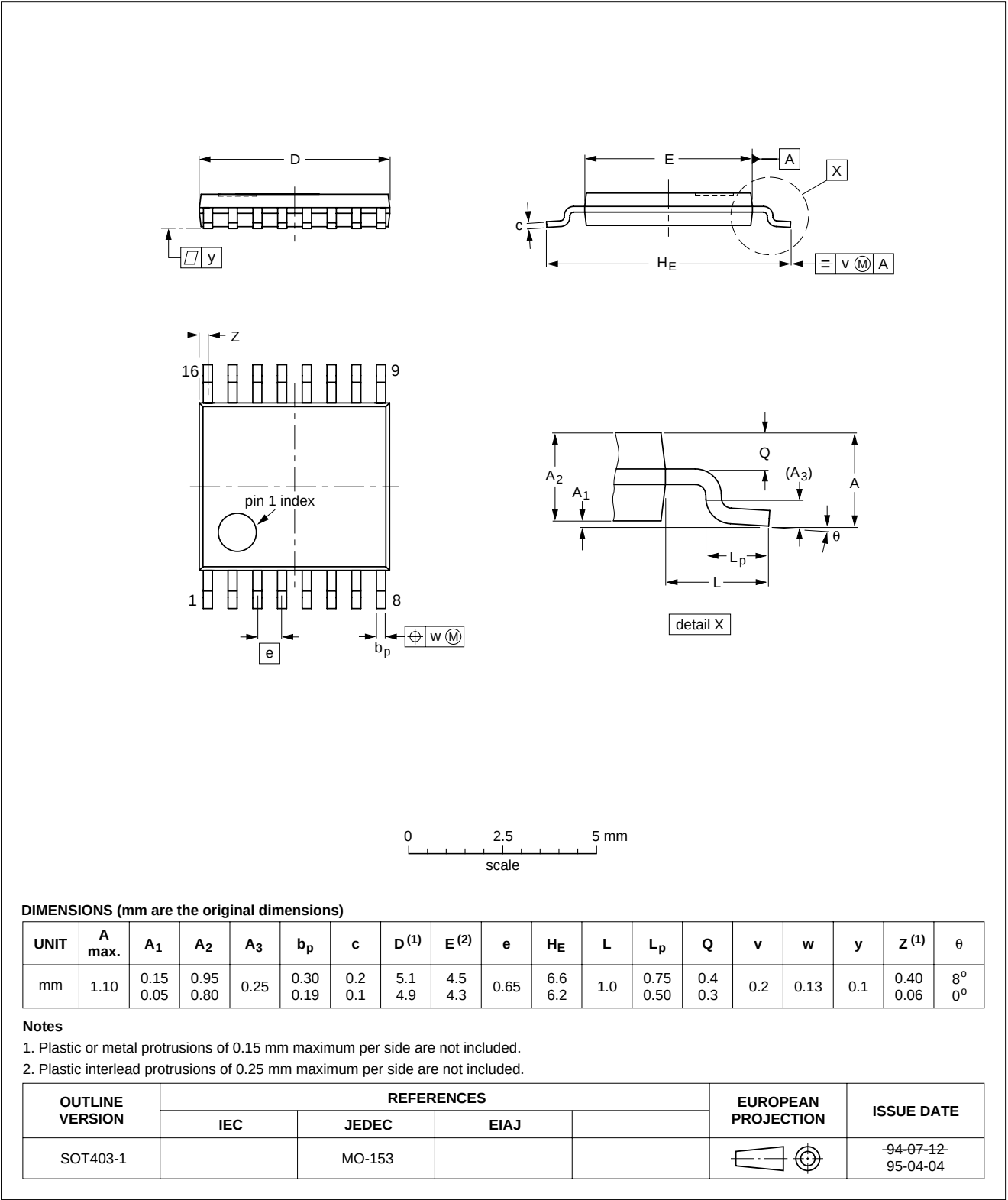


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TSSOP16: plastic thin shrink small outline package; 16 leads; body width 4.4 mm

SOT403-1



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SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"Data Handbook IC26; Integrated Circuit Packages"* (order code 9398 652 90011).

DIP

SOLDERING BY DIPPING OR BY WAVE

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg\ max}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

REPAIRING SOLDERED JOINTS

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

SO, SSOP and TSSOP

REFLOW SOLDERING

Reflow soldering techniques are suitable for all SO, SSOP and TSSOP packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method.

Typical reflow temperatures range from 215 to 250 °C. Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

WAVE SOLDERING

Wave soldering can be used for all SO packages. Wave soldering is **not** recommended for SSOP and TSSOP packages, because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering is used - **and cannot be avoided for SSOP and TSSOP packages** - the following conditions must be observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow and must incorporate solder thieves at the downstream end.

Even with these conditions:

- **Only consider wave soldering SSOP packages that have a body width of 4.4 mm, that is SSOP16 (SOT369-1) or SSOP20 (SOT266-1).**
- **Do not consider wave soldering TSSOP packages with 48 leads or more, that is TSSOP48 (SOT362-1) and TSSOP56 (SOT364-1).**

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.